Decentralized Static Output-Feedback Control via Networked Communicaiton

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American Control Conference, Montreal - FrB13.5 29 June, 2012

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Problem Description

We consider stabilizing decentralized control design for

- \blacktriangleright large-scale continuous-time linear plant
- \triangleright via a multi-purpose network with
	- shared communication: not all outputs and inputs can be communicated simultaneously
	- uncertain time-varying transmission intervals $h_k \in [h, \overline{h}]$ $\forall k \in \mathbb{N}$

Outline

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The plant is given by a set of disjoint subsystems:

$$
\mathcal{P}_i: \left\{ \begin{array}{rcl} \dot{x}_i(t) & = & A_i x_i(t) + B_i \hat{u}_i(t) + \sum_{\substack{j=1 \\ j \neq i}}^N \left(A_{i,j} x_j(t) + B_{i,j} \hat{u}_j(t) \right), \\ y_i(t) & = & C_i x_i(t) + \sum_{\substack{j=1 \\ j \neq i}}^N C_{i,j} x_j(t), \end{array} \right.
$$

which can be written as

$$
\mathcal{P}:\n\begin{cases}\n\dot{x}(t) = \underbrace{\begin{bmatrix}\nA_1 & \cdots & A_{1,N} \\
\vdots & \ddots & \vdots \\
A_{N,1} & \cdots & A_N\n\end{bmatrix}}_{\mathcal{C}}\n\end{bmatrix}\n\begin{bmatrix}\nx_1(t) \\
\vdots \\
x_N(t)\n\end{bmatrix} + \underbrace{\begin{bmatrix}\nB_1 & \cdots & B_{1,N} \\
\vdots & \ddots & \vdots \\
B_{N,1} & \cdots & B_N\n\end{bmatrix}}_{\mathcal{B}}\n\begin{bmatrix}\n\hat{u}_1(t) \\
\vdots \\
\hat{u}_N(t)\n\end{bmatrix}, \\
y(t) = \underbrace{\begin{bmatrix}\nC_1 & \cdots & C_{1,N} \\
\vdots & \ddots & \vdots \\
C_{N,1} & \cdots & C_N\n\end{bmatrix}}_{\mathcal{C}}\n\begin{bmatrix}\nx_1(t) \\
\vdots \\
x_N(t)\n\end{bmatrix},
$$

The set of subsystems can be expressed as

$$
\mathcal{P}(t) := \begin{cases} \dot{x}(t) & = & Ax(t) + B\hat{u}(t) \\ y(t) & = & Cx(t) \end{cases}
$$

We can express this system with time-varying transmission intervals in the following way

$$
\mathcal{P}_{h_k} := \left\{ \begin{array}{rcl} x_{k+1} &=& \bar{A}_{h_k} x_k + \bar{B}_{h_k} \hat{u}_k \\ y_k &=& C x_k \end{array} \right., \quad h_k \in [\underline{h}, \overline{h}]
$$

where $\bar{A}_{h_k}:=e^{Ah_k}$, $\bar{B}_{h_k}:=\int_0^{h_k} e^{As} ds B$

* Time-varying delays also can be incorporated similarly as an additive uncertainty

Model - Shared Communication Medium

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Shared Communication Medium

Only one node is allowed to transmit information at each transmission time

Node **- A collection of sensors and/or actuators are allowed** communicate over a network simultaneously

Periodic Protocol - grant network access to each node in a periodic fashion

 $\sigma_k \in \{1, 2, ..., \bar{N}\}$ denotes the node that has access at transmission time $k \in \mathbb{N}$
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Shared Communication Medium

$$
\hat{u}_{j,k} = \begin{cases} u_{j,k} & \text{if node } j \text{ has access} \\ \hat{u}_{j,k-1} & \text{otherwise} \end{cases}
$$

Shared Communication Medium

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$$

Mathematically we express this as

$$
\left[\begin{array}{c}\hat{u}_k\\\hat{y}_k\end{array}\right]=\Gamma_{\sigma_k}\left[\begin{array}{c}u_k\\y_k\end{array}\right]+(I-\Gamma_{\sigma_k})\left[\begin{array}{c}\hat{u}_{k-1}\\ \hat{y}_{k-1}\end{array}\right]
$$

where $\Gamma_{\sigma_{k}} = \text{diag}(\gamma_{j,\sigma_{k}})$

$$
\gamma_{i,\sigma_k} = \left\{ \begin{array}{ll} 1 & \text{if } u_{j,k}/y_{j,k} \text{ has network access} \\ 0 & \text{otherwise} \end{array} \right.
$$

Model - Overview

Plant Dynamics:

$$
\mathcal{P}_{h_k} := \left\{ \begin{array}{rcl} x_{k+1} &=& \bar{A}_{h_k} x_k + \bar{B}_{h_k} \hat{u}_k \\ y_k &=& \bar{C} x_k \end{array} \right. \qquad h_k \in [\underline{h}, \overline{h}]
$$

Shared Communication:

$$
\begin{cases} \n\hat{u}_k = \Gamma_{\sigma_k}^u u_k + (I - \Gamma_{\sigma_k}^u) \hat{u}_{k-1} \\
\hat{y}_k = \Gamma_{\sigma_k}^v y_k + (I - \Gamma_{\sigma_k}^v) \hat{y}_{k-1} \n\end{cases} \n\sigma_k \in \{1, ..., \bar{N}\}
$$

Controller:

$$
C_{\sigma_k}: u_k = K_{\sigma_k} \hat{y}_k
$$

$$
K_{\sigma_k} = diag(K_{\sigma_k,1}, K_{\sigma_k,2}, ..., K_{\sigma_k,N}),
$$

The closed loop model can be written as a discrete-time switched system with exponential uncertainty:

$$
\bar{X}_{k+1} = \tilde{A}_{h_k, \sigma_k} \bar{X}_k, \quad h_k \in [\underline{h}, \overline{h}], \ \sigma_k \in \{1, ..., N\}
$$

where

$$
\bar{\mathbf{x}}_k^{\top} = \begin{bmatrix} \mathbf{x}_k & \hat{\mathbf{u}}_{k-1} & \hat{\mathbf{y}}_{k-1} \end{bmatrix}
$$

and

$$
\tilde{A}_{h_k,\sigma_k} = \begin{bmatrix}\n\bar{A}_{h_k} + \bar{B}_{h_k} \Gamma^u_{\sigma_k} K_{\sigma_k} \Gamma^{\mathcal{Y}}_{\sigma_k} C & \bar{B}_{h_k} \Gamma^u_{\sigma_k} K_{\sigma_k} (I - \Gamma^{\mathcal{Y}}_{\sigma_k}) & \bar{B}_{h_k} (I - \Gamma^u_{\sigma_k}) \\
\Gamma^{\mathcal{Y}}_{\sigma_k} C & (I - \Gamma^{\mathcal{Y}}_{\sigma_k}) & 0 \\
\Gamma^u_{\sigma_k} K_{\sigma_k} \Gamma^{\mathcal{Y}}_{\sigma_k} C & \Gamma^u_{\sigma_k} K_{\sigma_k} (I - \Gamma^{\mathcal{Y}}_{\sigma_k}) & (I - \Gamma^u_{\sigma_k})\n\end{bmatrix}
$$

Structural Constraints

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Structural Constraints

-> Time-varying structural constraints

Design Problem:

Given a decomposition, a protocol, and $[\underline{h}, \overline{h}]$, how to choose $\Gamma^{u}_{\sigma_{k}} K_{\sigma_{k}}$ such that the closed-loop NCS is globally exponentially stable?

Goal:

Provide LMI conditions to design $\Gamma_{\sigma_k}^u$ \mathcal{K}_{σ_k} using the Lyapunov candidate

$$
V_{\sigma_k}(x_k) = \bar{x}_k^{\top} P_{\sigma_k} \bar{x}_k > 0, \quad \bar{x}_k \neq 0
$$

which must satisfy

$$
\Delta V_{\sigma_k}(x_k) = \bar{x}_k^{\top} (\tilde{A}_{h_k,\sigma_k}^{\top} P_{\sigma_{k+1}} \tilde{A}_{h_k,\sigma_k} - P_{\sigma_k}) \bar{x}_k < 0, \quad \bar{x}_k \neq 0
$$

Design

Challenges:

- 1. Uncertain nonlinearity
- 2. Time-varying structural constraints

Our Solution:

- 1. Create a convex overapproximation of the closed-loop system
- 2. Rewrite the constrained matrices as a multi-gain output feedback
- 3. Combine above two techniques to form sufficient LMI synthesis conditions

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Due to the transmission variance h_k ∈ $[h, h]$ $\forall k \in \mathbb{N}$, there is an infinite amount of sequences to check for guaranteeing stability

 $\left\{\tilde{A}_{h,\sigma} \mid h \in [\underline{h}, \overline{h}]\right\}$

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$$
\left\{\tilde{A}_{h,\sigma} \mid h \in [\underline{h}, \overline{h}]\right\} \subseteq \left\{\sum_{j=1}^{M} \alpha^{j} \left(F_{\sigma,j} + G_{j} \Delta H_{\sigma}\right)\right\}
$$

Therefore we synthesize controllers with an overapproximation of the original model, which is achieved by

(i) gridding a finite number of points in $[h, h]$ (ii) adding norm-bounded uncertainty to each grid point to capture the non-linearity between grid points.

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Therefore we synthesize controllers with an overapproximation of the original model, which is achieved by

(i) gridding a finite number of points in $[h, h]$ (ii) adding norm-bounded uncertainty to each grid point to capture the non-linearity between grid points.

- + introduces arbitrarily little conservatism
- $+$ direct control over the complexity of the overapproximation

[1] Donkers, et. al. Trans. Autom. Control 2011
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$$
\Gamma^{u} K = \begin{bmatrix} \cdot & \cdot & \cdot & \cdot \\ \cdot &
$$

$$
\bar{x}_{k+1} = \left[\hat{A}_{h_k,\sigma_k} + \sum_{i=1}^N (\hat{B}_{h_k,\sigma_k,i} \bar{K}_{\sigma_k,i} \hat{E}_{\sigma_k,i})\right] \bar{x}_k
$$

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If there exist matrices such that

1. For
$$
j \in \{1, ..., \tilde{N}\}, m \in \{1, ..., M\}
$$

\n
$$
\begin{bmatrix}\nG_{\sigma_j} + G_{\sigma_j}^\top - P_j & \Xi_1(j, m)^\top & 0 & \Xi_2(j)^\top \\
\star & P_{j+1} & g_m R_{j,m} & 0 \\
\star & \star & \star & R_{j,m} & 0 \\
\star & \star & \star & R_{j,m}\n\end{bmatrix} \succ 0
$$

$$
\begin{aligned} \Xi_1(j,m) &:= \mathcal{A}_{\sigma_j,m} G_{\sigma_j} + \sum_{i=1}^N \mathcal{B}_{\sigma_j,m,i} Z_{\sigma_j,i} C_{\sigma_j,i} \\ \Xi_2(j) &:= \mathcal{D}_{\sigma_j} G_{\sigma_j} + \sum_{i=1}^N \mathcal{E}_{\sigma_j,i} Z_{\sigma_j,i} C_{\sigma_j,i} \end{aligned}
$$

2. For $l \in L_{y, i}$, $i \in \{1, ..., N\}$

$$
X_{l,i}e_{l,i}=e_{l,i}G_l
$$

then $\bar{K}_{l,i} = Z_{l,i}X_{l,i}^{-1}$ renders the closed-loop GES.

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Numerical Example

Plant Model

Communication Protocol (when shared):

$$
\Gamma_1^u = diag(1, 1, 0, 0), \qquad \Gamma_2^u = diag(0, 0, 1, 1),
$$

\n
$$
\Gamma_1^y = diag(1, 1, 1, 0, 0, 0), \qquad \Gamma_2^y = diag(0, 0, 0, 1, 1, 1),
$$

\n
$$
\sigma_k = 1, 2, 1,
$$

Coupling Visualization

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Numerical Example

 $h_k \in [(1-\delta)h_\star, (1+\delta)h_\star]$

Average computation time to solve LMI: C1, C2: 15 seconds C3: 40 seconds

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Conclusions

- \triangleright We presented a model for an NCS which includes
	- varying transmission intervals $[h, h]$
	- shared communication medium (protocol)
	- The controllers are
		- static
		- decentralized
		- switch based on protocol
- \triangleright Sufficient conditions for controller synthesis were provided for the
	- decentralized problem setting
	- NCS problem setting
	- unification of the above two problem settings
	- + Single-shot design ('a posteriori' analysis not required)
	- No structure was imposed on the Lyapunov function
	- + Extendable to synthesizing decentralized observer-based controllers

